













anticipated, dissimilar IR responses were observed per location and therefore reasonable to predict similar behaviors for the  $SNR_{Rcv}^0$  responses at those locations. This is corroborated subsequently.

**4.1.2 Received SNR,  $SNR_{Rcv}^0$ , measurement:** Following the immediately outlined channel IR measurements, the measured  $SNR_{Rcv}^0$  responses distributed over the passenger plane grid was completed and shown in Fig. 7. A steady background noise (includes

**Fig. 7:** Measured  $SNR_{Rcv}^0(x, y)$  in-car.

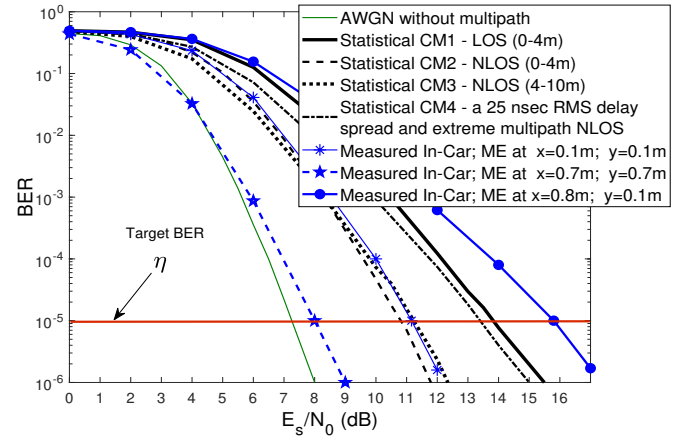
noise floor) of -91 dBm was observed in-car and rapid fluctuations apparent in the passenger plane (Fig. 7). This is in line with the dissimilar IRs of Section 4.1.1, is due to the rich multipath behavior inside the car body and is corroborated by the multiple echoes of the direct path with fading (path-loss attenuation) observed in Fig. 6. The statistical properties of the measured  $SNR_{Rcv}^0$  data were: maximum 14.1 dB, minimum 8.1 dB, mean 11.1 dB, and standard deviation 0.16 dB. In addition, the highly fluctuated  $SNR_{Rcv}^0$  of Fig. 7 does not assure an equal target BER over the passenger plane grid and justifies the need of measuring the entire grid plan ( $N$  points); this is because the BER response will also fluctuate and differ from the  $\eta$ . Since the signal strength is dependent on the antenna efficiency, the  $SNR_{Rcv}^0$  can be improved and meet the target BER  $\eta$  by redefining the radiation pattern of the access point antenna, next.

## 4.2 Stage-2: The BER and the $SF_d$

In this section, the BER and the desired antenna space-factor ( $SF_d$ ) performance in-car is evaluated. The adapted UWB MB-OFDM model presented in Section 3.2 is used and the UWB channel custom block redefined using a friendly graphical user interface (GUI) with setting parameters that considered the system behavior in the following communication channels, statistical AWGN, CM1-4, and the channel measurements reported in Section 4.1. The evaluation process comprises two phases, 1) the predicted BER performance using the earlier experimentally measured in-car channel and 2) the  $SF_d$  for improved BER responses in-car.

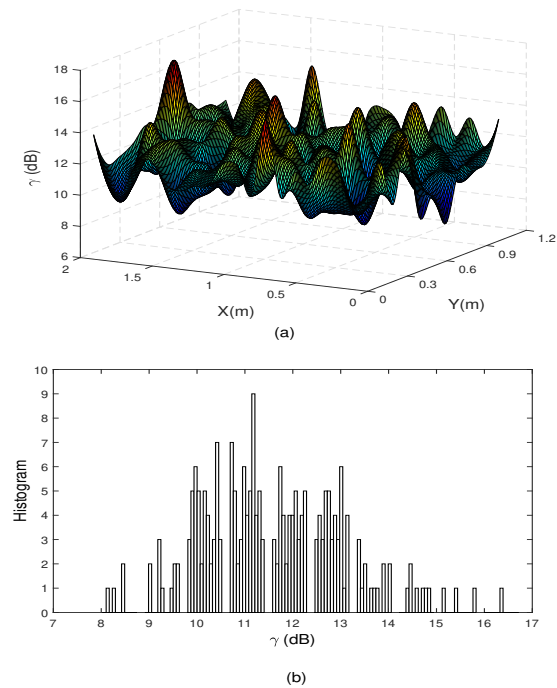
**4.2.1 The predicted BER vs.  $E_s/N_0$  performance using the measured  $h^0(t, x, y)$  and  $SNR_{Rcv}^0$ :** Essentially, how the in-vehicle environment affected the channel model is shown in Fig. 8 where a deviation in the  $E_s/N_0$  responses compared to other statistical models, CM1-4, was observed. The MB-OFDM performance was initially characterized in an AWGN channel to obtain the minimum SNR for target BER  $\eta$  and helped quantifying the effect of channel dispersion; the SNR was found to be  $\sim 7.4$  dB. Dissimilar SNR-BERs in each mobile equipment (ME) location encouraged measuring the multiple locations in-car and established the target SNR  $\gamma$  (where the BER crosses  $\eta$ ) for 220 measured mobile equipment locations in-car; only 3 locations are shown for clarity. Results corroborate the predictions presented in Section 4.1.2, where dissimilar BER performances were anticipated in the passenger grid plan ( $N$  points). This justifies the use of the 220 readings (a high number means accuracy) and the practical on-site measurements

overcome unrealistic statistical channels in this application - unrealistic because these canonical form channels do not fully describe the in-car application presented in this manuscript. The slightly improved performance on CM4 over the CM1 (LOS) was due to the car body acting as a reverberation channel which favored from the collection of multipaths.



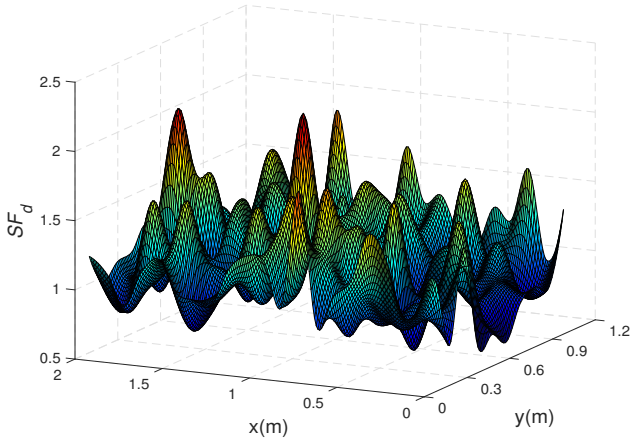
**Fig. 8:** BER vs.  $E_s/N_0$  using statistical and measured channels with different IRs.

**4.2.2 The  $SF_d(x, y)$  that attains the target BER performance in the passenger plan ( $N$  points):** For the assessment,  $\gamma(x, y)$  is calculated using (4) and  $\gamma = E_s/N_0 \times (f_s/B)$ , where  $f_s$  is the channel data rate and  $B$  the bandwidth. The result is plotted in Fig. 9(a) and shows a  $\gamma(x, y)$  as dictated by the  $\eta = 10^{-5}$  (Fig. 8). A histogram, presented in Fig. 9(b), shows the highly varied  $\gamma$  responses over the passenger plane grid, corroborates the need of measuring  $N$  points, and prevails over the predominantly use of statistical channel models for their lack of accuracy (scarce number of registered data points [17–23]) in this application.



**Fig. 9:** (a) The calculated  $\gamma(x, y)$  in-car, and (b) the histogram of the  $\gamma$ .

Following the calculated  $\gamma(x, y)$ , the  $SF_d(x, y)$  was found. For the calculation, (8), evolving the  $SNR_{Rcv}^0$  (Fig. 7) and the  $\gamma$  (Fig. 9a), was used. The result is plotted in Fig. 10 and used to optimize the physical aperture of the prospective antenna and its associated continuous source distribution, next.



**Fig. 10:** The calculated desired antenna space-factor  $SF_d(x, y)$  in-car.

#### 4.3 Stage-3: Radiation pattern-and-source synthesis

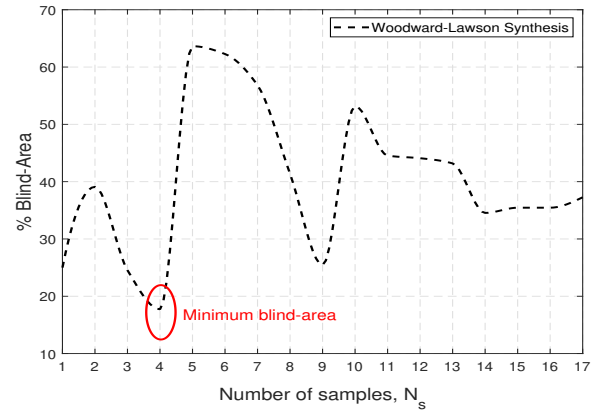
The physical aperture and associated continuous source distribution of the prospective antenna is now determined. Particularly, the Woodward-Lawson pattern synthesis technique introduced in Section 3.3.2 and the  $SF_d$  response presented in Fig. 10, is used to determine the required continuous source,  $I_{fin}(x', y')$  and  $\Phi_{fin}(x', y')$ . The  $SF_d$  was obtained using 220  $N$  points since a higher number of samples in (20) yields to a highly complex inverse operation to matrix (16). Hence, we limited the number of samples to  $N_s$ ,  $\rightarrow N_s \leq N$ . Due to the limitations of using  $N_s$  (rather than  $N$ ), the BER adjusts to  $BER^{fin}$  and hence to  $SF_{fin}$ . Since  $N_s$  defines the physical aperture of the antenna, Fig. 4, whose dimensions  $l_x = l_y = \lceil \sqrt{N_s} \rceil \lambda$ , [37], the lower the  $N_s$ , the smaller the aperture (preferred). But, as  $N_s$  impacts on the BER performance (later corroborated in this Section), we established herein a prospective antenna design whose  $N_s$  is the optimization parameter to minimize the blind spots (1) in-car.

To find the continuous source (18) required by the antenna we rearranged the initial goal (2) such that the optimization problem becomes:

$$\begin{aligned} & \text{Minimize} \quad \% \text{Blind Area (1)} \\ & \quad \quad \quad N_s \\ & \text{subject to} \quad N_s \leq N, \end{aligned} \quad (22)$$

To determine the appropriate  $N_s$  value, we close looked at the individual  $N = 220$  samples of the  $SNR_{Rcv}^0$  (Fig. 7) and selected those samples with deprived SNR; the selected samples correspond to those deeply faded responses in the passenger grid plan. To improve the received signals in-car and the BER consequently, we computed the minimum blind-area solution probable and the continuous source coefficients for that solution, next.

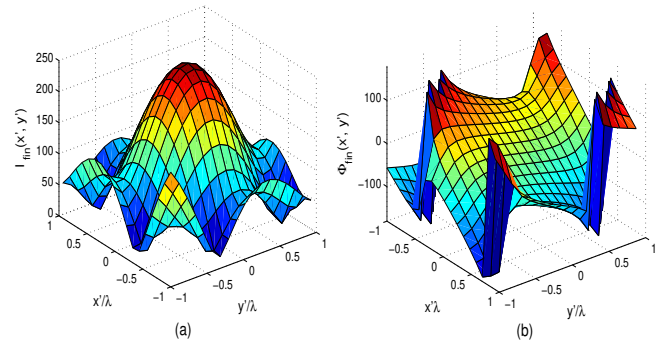
**4.3.1 The optimization of the blind-areas in-car:** We used (22) and the Woodward-Lawson synthesis of Section 3.3.2 for the purpose and plot aided by Matlab the percentage blind-area in regards to  $N_s$  in Fig. 11. The response is non-monotonic with  $N_s$  settled to the adopted scheme and not to a synthesized antenna size. By observation, the overall blind-area is  $N_s$  dependent and severely varies upon  $N_s$ . Furthermore, values of  $N_s > 17$  have no real solution since the inverse matrix calculation of the invertible square matrix (16) is singular (not feasible) to working precision. Therefore



**Fig. 11:** The minimum blind-area in-car used for optimization.

$N_s$  is limited to  $1 \leq N_s \leq 17$  and the optimized solution to (22) found to be  $N_s = 4$  (the lowest blind-area in Fig. 11). This is in fact the 4 lowest SNRs in the passenger grid plan and will be demonstrated, later in Section 5, to be an optimal value for the improved BER performances in-car.

**4.3.2 The required continuous source coefficients:** We used the immediately reported  $N_s = 4$ , which in fact corresponds to 4 optimal samples of the  $SF_d$  (Fig. 10). Since the continuous source coefficients (amplitude and phases) can be found from the  $SF_d$  (by finding  $[b_n]$ ), Section 3.3.2, hence, we use the inverse pattern synthesis technique to retrieve every  $I_{fin}(x', y')$  and  $\Phi_{fin}(x', y')$  composing the prospective antenna's physical aperture. The result is depicted in Fig. 12, and leads to an antenna (Fig. 4) of dimensions  $l_x = l_y = \lceil \sqrt{N_s} \rceil \lambda = 2\lambda$ . Using the antenna's physical aperture as



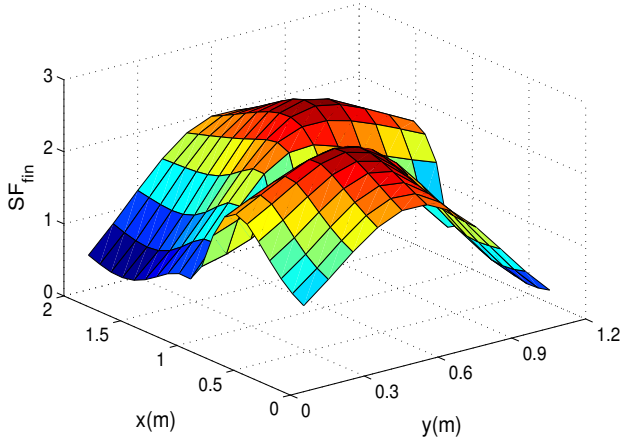
**Fig. 12:** Required continuous source coefficients (a)  $I_{fin}(x', y')$  (b)  $\Phi_{fin}(x', y')$  showing the prospective antenna's physical aperture in regards to wavelength,  $\lambda$ .

a contribution of the continuous source coefficients immediately outlined (shown in Fig. 12), the  $SF_{fin}(x, y)$  of the prospective antenna is calculated and the result depicted in Fig. 13. Primarily given by the computational limitations of  $N_s \leq N$  and the inherent antenna's aperture dimensions, the  $SF_{fin}$  differed from the  $SF_d$ , but did not present inhibition for improving the UWB MB-OFDM communications in-car. This is corroborated in the results section, subsequently.

## 5 Results

The improved in-vehicle UWB MB-OFDM communications is evidenced from the improved responses (BER and blind-area) obtained when using the prospective antenna compared to the standard and is detailed next.

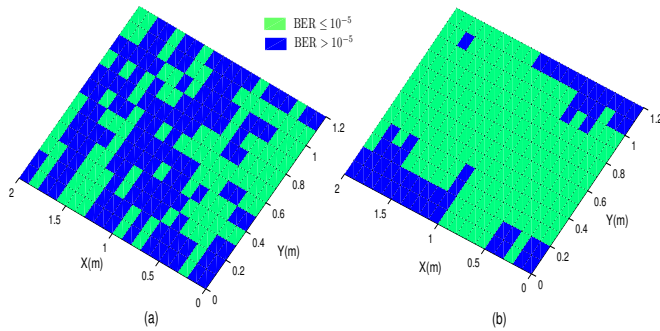




**Fig. 13:** The calculated  $SF_{fin}(x, y)$  in-car using the prospective antenna as access point.

### 5.1 The improved BER

Using the adapted Simulink model presented in Section 3, the BER performance in-car is predicted when using the standard antenna and the prospective antenna for the access point. Comparison results show the  $BER^0(x, y)$  vs. the  $BER^{fin}(x, y)$ , in Fig. 14(a) and (b) respectively, distributed over the passenger grid plan in a color map surface plot form. By observation, the BER performance in-



**Fig. 14:** The predicted BER in-car (a)  $BER^0$  using the standard antenna (b)  $BER^{fin}$  using the prospective antenna

car is outperformed for the case where the prospective antenna is used since a far inferior number of blind-areas ( $BER > 10^{-5}$ ) are apparent. The blind areas still present in Fig. 14(b) depended on the selection of the  $N_s$  samples that were used for optimization. This value was limited upon the antenna realization (realistic size) to 4 (Section 4.3) and led to few error rates in certain points of the grid plan. Yet, a largely improved error rate is seen as a whole. The accomplished  $BER^{fin}(x, y)$  is therefore principally achieved as a result of the well-defined continuous source coefficients,  $I_{fin}(x', y')$  and  $\Phi_{fin}(x', y')$  of Fig. 12.

### 5.2 The improved blind-area

Using the results of Fig. 14, the percentage blind-area was computed for comparison between the responses of the prospective antenna vs. those of the standard when used as access point. The comparison is provided in Table 1 and the result achieved by the prospective antenna reduced the %Blind-Area (1) in 37.73% compared to that of the standard antenna. This translates into high-efficient (capacity and data rate) UWB-OFDM communications in-car.

**Table 1** Performance comparison when using the prospective antenna vs. the standard antenna for the access point.

	% Blind Area
Standard Antenna	55.45%
prospective Antenna	17.72%

### 5.3 Feasible antenna solution

The prospective antenna gained from this research contribution is given by continuous source coefficients (amplitudes and phases) composing the antenna's physical aperture. The fabrication of this feasible antenna is suggested as future research. Although the rectangular-like planar aperture of the antenna ( $l_x = l_y = 2\lambda$ ) was found tolerable for the in-car application, to exactly reflect the  $I_{fin}(x', y')$  and  $\Phi_{fin}(x', y')$  on a continuous antenna aperture (i.e: a patch) needs to be addressed, but seems to be a better solution than an array-made antenna since the feeding network of the array would be large and lossy. The antenna-array becomes large since multiple source coefficients would be required to form the array elements. A discretization of the continuous source (e.g., discrete source) can be used to estimate the radiation pattern of the antenna close to the  $SF_d$  [37]. The accuracy of this radiation pattern depends on the sampling occurrence, the permissible difference between the  $SF_d$  and the approximate pattern, and the  $BER_{fin}$ . The sampling can be met using the Root-Matching and perturbation techniques [40]; all have their own limitations over accuracy. Prototypes based on dielectric-made domes for shaping the antenna's radiation pattern [41] are suggested for investigation using the hereby proposed channel-based antenna synthesis; that would bring antennas with spatially selective gains, ideal for the antenna solution.

## 6 Conclusion

A channel-based antenna synthesis for improved in-vehicle UWB MB-OFDM communications has been proposed and presented. The synthesis allows for optimizing an antenna design for the scenario. The radiation pattern (an optimized SF) of the antenna (given by continuous source coefficients and a rectangular-like planar aperture) was modeled to provide improved in-vehicle UWB communications. This was achieved by an antenna whose SF provided (uniform) SNR-BER response in the passenger plane in-car and attributed to lowering the blind spots in the plane. Although this synthesis can be applicable to other vehicles, the extrapolation of the contemporary SF is limited to cars of similar dimensions since the IRs of the channel would vary. Any variation in the channel might not significantly influence on the primary antenna selection, but if a fully customized antenna is desired a new set of measurements should be performed. To corroborate the improved BER performance in-car, the results obtained using the prospective antenna were compared to those of a standard antenna. A new set of in-car channel measurements was performed, using the standard antenna, to overcome the unrealistic behavior of existing channel models in this scenario. The BERs were predicted using this genuine channel and provided the basis for estimating the SF of the access point antenna. The use of the prospective antenna showed an alleviated blind-area performance compared to that using the standard antenna and supports the optimized design (SF + continuous source coefficients + rectangular-like planar aperture) as a candidate for the access point in high-efficient UWB-OFDM communications in-car.

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